

This application note describes crest factor reduction and an Altera® crest factor reduction solution.

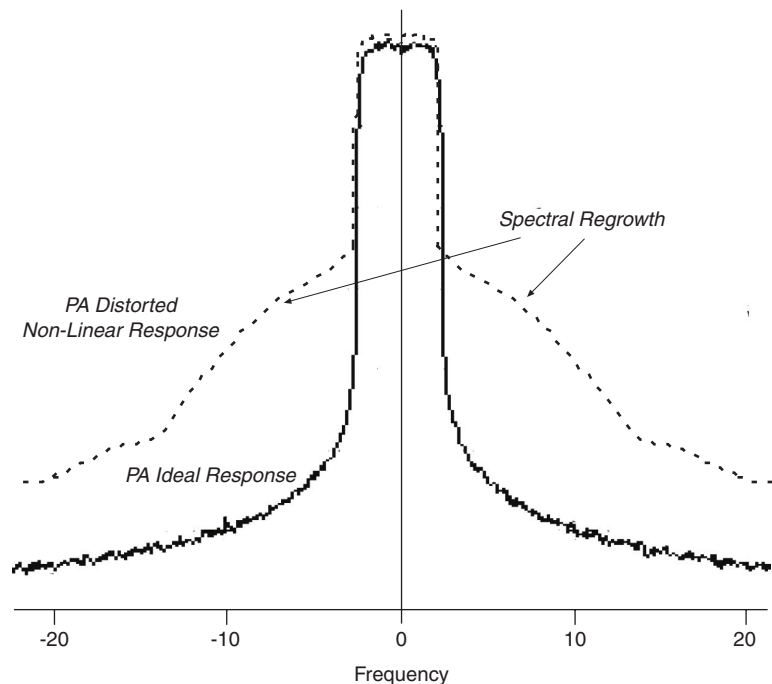
## Overview

A high peak-to-mean power ratio causes the following effects:

- In-band distortion:
  - High error vector magnitude (EVM)
  - Degrades performance at receiver
- Out-of-band distortion:
  - Increased adjacent channel leakage ratio (ACLR)
  - Degrades performance of users in adjacent channels

Figure 1 shows spectral regrowth with power amplifier (PA) distortion.

**Figure 1. Spectral Regrowth with PA Distortion**



Crest factor reduction (CFR) reduces the output peak-to-average ratio by clipping and allows additional gain on the output of CFR. You can operate closer to the amplifier compression point, therefore it is more efficient. Figure 2 shows output clipping with CFR.

**Figure 2. Output Clipping with CFR**

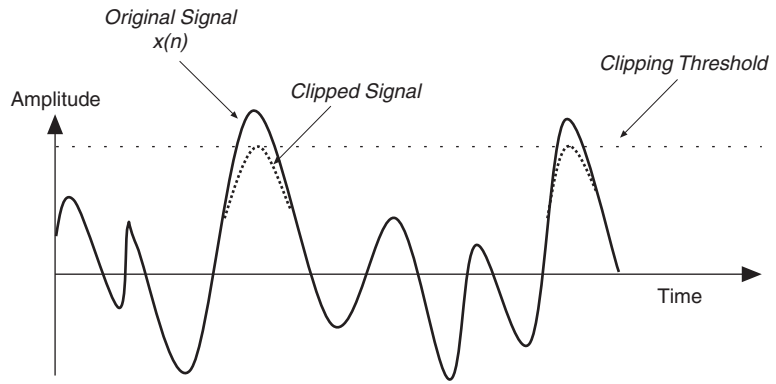
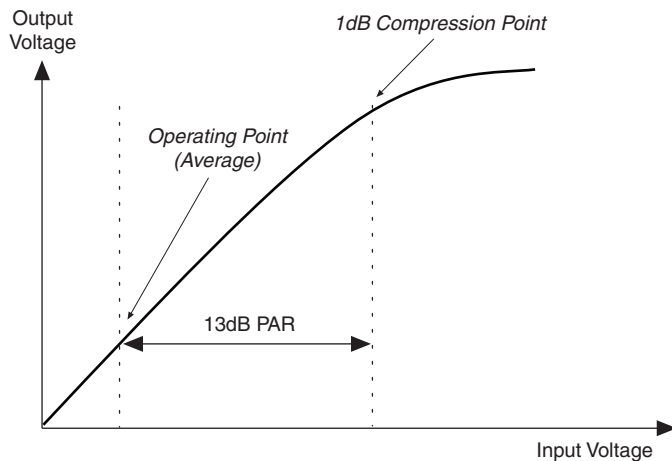
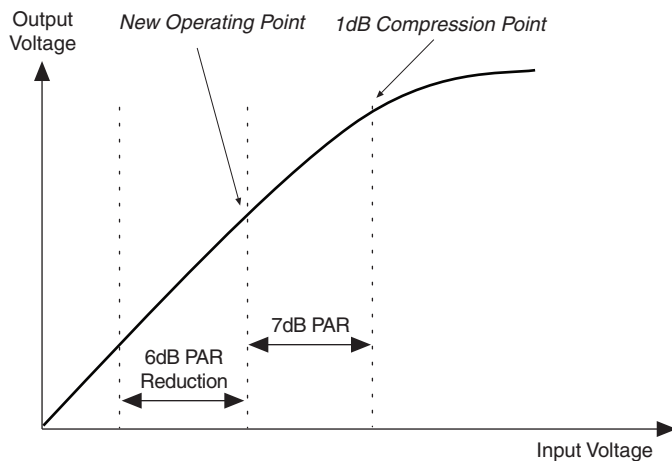


Figure 3 shows the amplifier transfer characteristics (before CFR); Figure 4 shows the amplifier transfer characteristics (after CFR).

**Figure 3. Amplifier Transfer Characteristic—Before CFR**

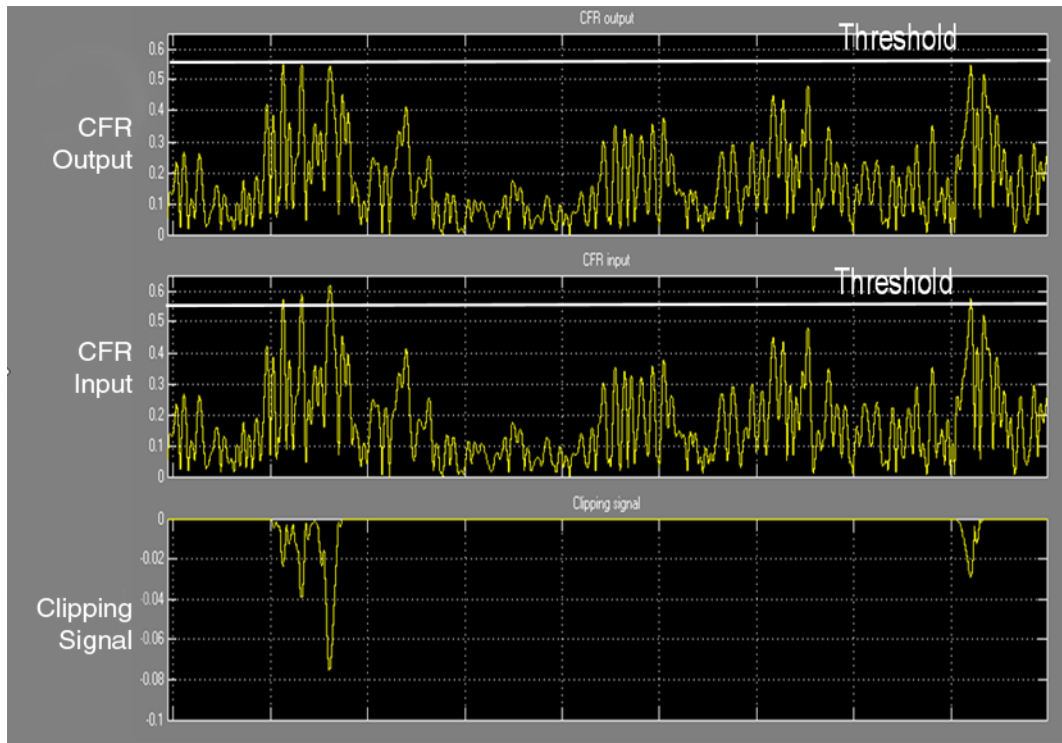


**Figure 4. Amplifier Transfer Characteristic—After CFR**

There are the following possible CFR techniques:

- Adaptive baseband:
  - CFR done separately on each carrier before upconversion
  - Measurements taken at interpolated baseband to determine which delayed baseband symbols to modify
  - Good for ACLR; not so good for peak-to-average ratio (PAR) reduction
- Intermediate frequency (IF) clipping and filtering
  - Hard clip the IF signal and filter the result to improve the ACLR
  - Works on composite IF signal
  - Can have poor ACLR; filters can be complex to implement
- Peak windowing technique
  - Filter clips signal rather than actual data
  - Better ACLR than IF clipping and filtering; can require large filters

Figure 5 shows the PAR reduction with peak windowing.

**Figure 5. Peak Windowing CFR Block Diagram**

The peak windowing technique compares the magnitude of the input signal ( $R_{IF}$ ) with the clipping threshold to calculate the clipping ratio  $C(n)$ . The filter clipping ratio helps the ACLR at the cost of EVM. You clip the delayed input signal with the filtered clipping ratio  $B(n)$ . With the peak windowing technique, the filter coefficients remain same for both single and multi carrier case, but it can require large filters.

## Functional Description

Figure 6 shows the CFR module.

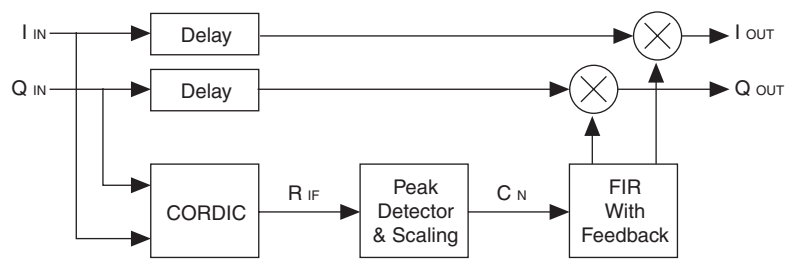
**Figure 6. CFR Module**

Figure 7 shows the CFR module in DSP Builder.

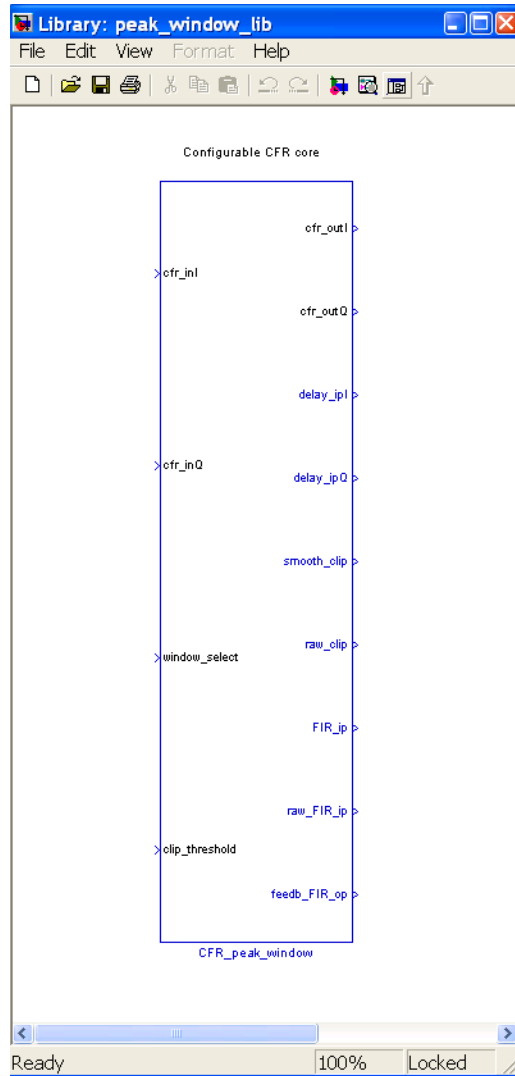
*Figure 7. CFR Module in DSP Builder*

Table 1 shows the module signals.

<b>Table 1. Signals (Part 1 of 2)</b>			
<b>DSP Builder Name</b>	<b>VHDL Name</b>	<b>Direction</b>	<b>Description</b>
cfr_inI	iInput_Ils	Input	Input I data. Input format is in the Q1.15 format, but can be changed in the <b>Simulation Parameters Setting</b> box.
cfr_inQ	iInput_Qls	Input	Input Q data.
window_select	iwin_selects	Input	Run-time configurable parameter to change the feedforward and feedback windows to either triangular ( <code>win_select = 0</code> ) or rounded ( <code>win_select=1</code> ). Window length is currently only configurable at synthesis, and can be changed <b>Simulation Parameters Setting</b> box.
clip_threshold	iInput_clip_vals	Input	Run-time configurable parameter, governing the clip threshold (set to 0.65 for an input normalized to 1.0). The actual (floating) value is scaled by the CORDIC gain (=1.64676) and converted to a Q3.26 number.
N/A	clock	Input	Clock. This signal does not appear on the DSP Builder entity, but is part of the VHDL.
N/A	sclrnp	Input	Synchronous clear. This signal does not appear on the DSP Builder entity, but is part of the VHDL.
cfr_outI (1)	oOutput_Clipped_I s	Output	Output I value, in the same number format as the input.
cfr_outQ (1)	oOutput_Clipped_Q s	Output	Output Q value.
delay_ipI (1)	oOutput_del_Is	Output	Input data delayed by the same amount as the CFR delay, which allows a comparison of input and output signals, and the impact of clipping.
delay_ipQ (1)	oOutput_del_Qs	Output	Input data delayed by the same amount as the CFR delay, which allows a comparison of input and output signals, and the impact of clipping.
smooth_clip (1)	oOutput_Smooth_Clips	Output	The clipping signal smoothed by the feedforward and feedback signals.

**Table 1. Signals (Part 2 of 2)**

DSP Builder Name	VHDL Name	Direction	Description
raw_clip (1)	oOutput_raw_clips	Output	The raw clipping signal. The positive part of the input minus the clipping threshold.
raw_FIR_ip (1)	oOutput_raw_FIR_ips	Output	The input to the feedforward filter (equal to the raw clipping signal minus the feedback filter).
FIR_ip (1)	oOutput_mFIR_ips	Output	The input to the feedforward filter (equal to the raw clipping signal minus the feedback filter) but with any negative values set to zero.
feedb_FIR_op (1)	oOutput_fFIR_ops	Output	The output of the feedback FIR, which reduces any overclipping caused by adjacent, high-clipping values.

**Note to Table 1:**

(1) Debug outputs, which are included for testing only, and if unconnected the Quartus II software synthesizes away any associated logic.

## Getting Started

This section involves the following steps:

- “Software Requirements” on page 8
- “Simulink Simulation” on page 8
- “Autogenerate RTL” on page 20
- “Synthesize the Design” on page 20
- “Perform RTL Simulation” on page 21
- “Verify Simulink Model & RTL” on page 25

## Software Requirements

This application note requires the following software:

- MATLAB version release 14
- Simulink version 6
- DSP Builder version 7.1
- The Quartus® II software version 7.1
- The ModelSim® simulation tool version 5.7d

## Simulink Simulation

To simulate in Simulink, follow these steps:

- “Set Up” on page 9
- “Parameterize the Design” on page 9



- “Run the Simulation” on page 17

### *Set Up*

To simulate in Simulink, follow these steps:

1. Open the MATLAB software.
2. Include additional paths:
  - a. Go to the top-level directory `<install directory>\altera_wireless`.
  - b. Type `add_wireless_paths`.

### *Parameterize the Design*

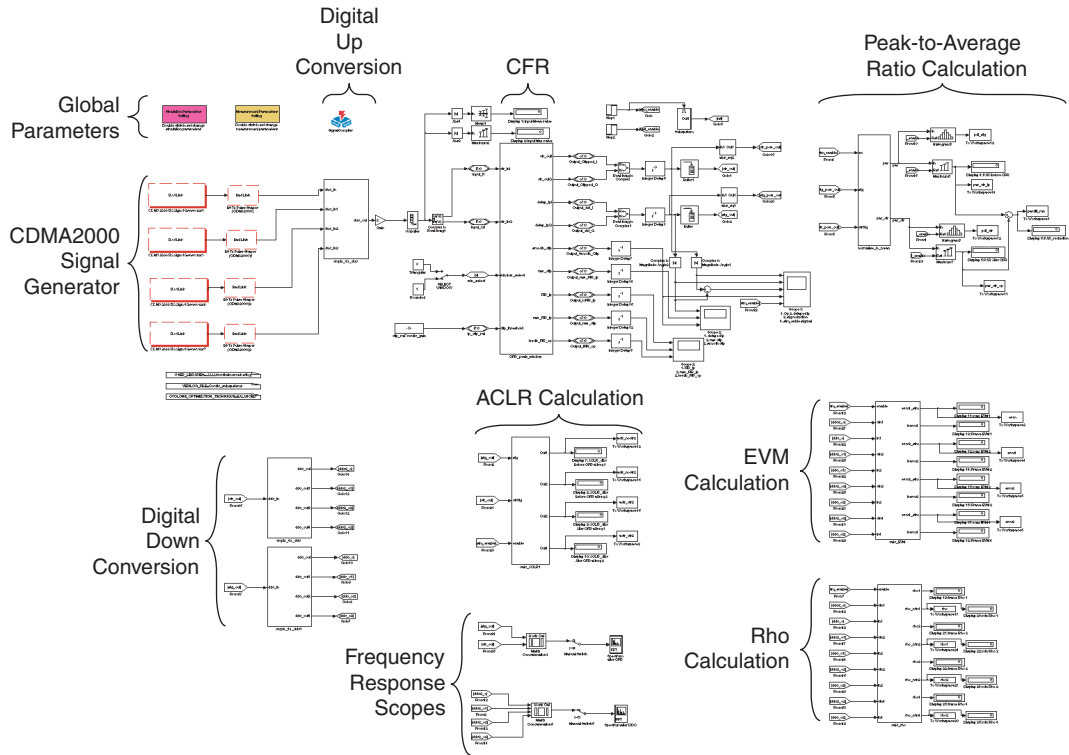
You can either use the two testbenches included with the reference design or instantiate the CFR in a new design (see “[Instantiate the CFR Module in a New Design](#)” on page 15). The two testbenches include an instantiated and parameterized CFR module, which you can use to explore the design.

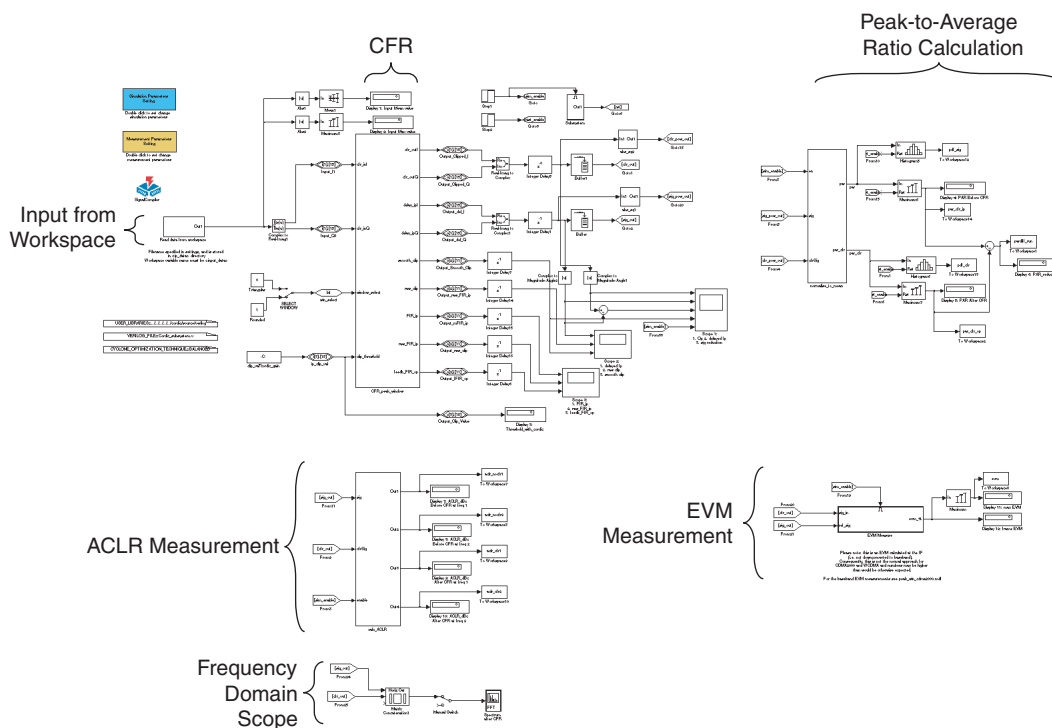
### **Testbenches**

Altera provide the following two testbenches:

- **Peak\_win\_cdma2000** (see [Figure 8](#)). The CDMA2000 source within testbench generates input data.
- **Peak\_win\_workspace** (see [Figure 9](#)). Input data is read from the workspace. A sample input file is included in the `ip_data` directory. This file (**FA4\_CDMA2000**) is for CDMA2000 with four carriers. The file length is 5 ms for an input data rate of 78.6 MHz (CDMA2000 rate upsampled by 64). With a pre-generated workspace variable as an input, the simulation time is shorter.

Figure 8. Peak\_win\_cdma2000 Testbench





- ✓ Open the model by typing `open_cfr`. This script performs the following actions:
  - Changes the directory to `<install directory>\altera_wireless\ip\cfr\test\matlab`.
  - Opens the **peak\_win\_cdma2000.mdl** testbench.
  - Updates the model to include the currently defined settings.



If you change the settings, or open a model directly in Simulink, you must update the model to reflect the variable settings:



Alternatively, open **testbench\_peak\_win\_wkspc** from the same location.

To parameterize the design double-click the global configuration parameter box on the top left of the design. [Figure 10](#) shows the Simulation Parameter Settings.

**Figure 10. Simulation Parameter Settings**

**Block Parameters: Double click to set change sim...** [?] [X]

Simulation Parameters Settings (mask)

Values set here are automatically propagated to the subsystems defined in this top level test bench.

Parameters

Clipping Amplitude Threshold (CFR block):  
0.65

IQ Signed Fractional: [number of bits].[ ]  
1

IQ Signed Fractional: [ ]. [number of bits]  
15

Clip Value Signed Fractional: [2]. [number of bits]  
11

Peak Window Length  
64

Chiprate:  
1.2288e6

CDMA2000 Frame length (seconds)  
[(frame\_length/chip\_rate)

BS Tx MATLAB Frame length (number of chips)  
2048

Carrier 1 Random Generator Seed:  
34

Carrier 2 Random Generator Seed:  
35

Carrier 3 Random Generator Seed:  
36

Carrier 4 Random Generator Seed:  
37

Pulse Shaper Group Delay (number of samples)  
24

Pulse Shaper interpolation factor:  
4

DUC interpolation factor:  
16

CFR latency (MATLAB frames): 2

DUC latency (MATLAB frames): 4

Carrier 1 frequency (Hz):  
5e6

Carrier 2 frequency (Hz):  
6.25e6

Carrier 3 frequency (Hz):  
7.5e6

Carrier 4 frequency (Hz):  
8.75e6

OK Cancel Help Apply

Table 4 shows the parameters for **peak\_win\_cdma2000**.



The **peak\_win\_workspace** testbench uses a subset of these parameters.

**Table 2. Parameters—peak\_win\_cdma2000 (Part 1 of 2)**

Parameter	Description
Clipping Amplitude (CFR block)	The value at which the clipping occurs, in floating point.
IO Signed Fractional: [number of bit].[]	The number of integer bits in the input data and data path in the design. For example, for one 16-bit signed number less than 1.0 (i.e. Q1.15), this is set as 1
IO Signed Fractional: [].[number of bit]	The number of fractional bits in the input data and data path in the design. For example, for one 16-bit signed number less than 1.0 (Q1.15), this is set as 15.
Clip Value Signed Fractional: [2].[number of bit]	The number of fractional bits for the clipping calculation in the design. Two magnitude and sign bits are used by default (the number is $2.0 > \text{value} \geq -2.0$ ).  Typically this number can be less than the input bit width, to reduce complexity. For example, 11 bits can be used here, making the overall number format Q2.11.
Peak Window Length	The width of the feed-forward filter impulse response, in samples. The feedback filter impulse response is one-half of this value.
Chip Rate	CDMA2000 chip rate (default setting is 1.2288 MHz).
CDMA2000 Frame length (seconds)	CDMA2000 frame length (default setting is 2,048 chips).
BS Tx MATLAB Frame Length (number of chips)	Processing interval size for the MATLAB/Simulink simulation (default setting is 2,048 chips).
Carrier 1 Random Generator Seed	The random seed from the CDMA2000 signal source.
Carrier 2 Random Generator Seed	
Carrier 3 Random Generator Seed	
Carrier 4 Random Generator Seed	
Pulse Shaper Group Delay (number of samples)	Latency associated with the root-raised cosine filter that pulse shapes the transmitted signal (default value is 24).
Pulse Shaper Interpolation Factor	Interpolation rate for pulse shaping filter (default value is 4, corresponding to $4 \times 1.2288 \text{ MHz} = 4.91 \text{ MHz}$ ).
DUC Interpolation Factor	Digital upconversion interpolation factor (default value is 16, corresponding to $16 \times 4.91 \text{ MHz} = 78.64 \text{ MHz}$ ).
CFR Latency (MATLAB frames)	Defines the time before measurements are taken in the simulation.

**Table 2. Parameters—peak\_win\_cdma2000 (Part 2 of 2)**

Parameter	Description
DUC Latency (MATLAB frames)	Defines the time before measurements are taken in the simulation.
Carrier 1 frequency (Hz)	The IF frequency of carrier 1.
Carrier 2 frequency (Hz)	The IF frequency of carrier 2.
Carrier 3 frequency (Hz)	The IF frequency of carrier 3.
Carrier 4 frequency (Hz)	The IF frequency of carrier 4.

Figure 9 shows the measurement parameters of the design. These parameters are specifically associated with how the testbench presents the results and do not affect the operation of the CFR block directly.

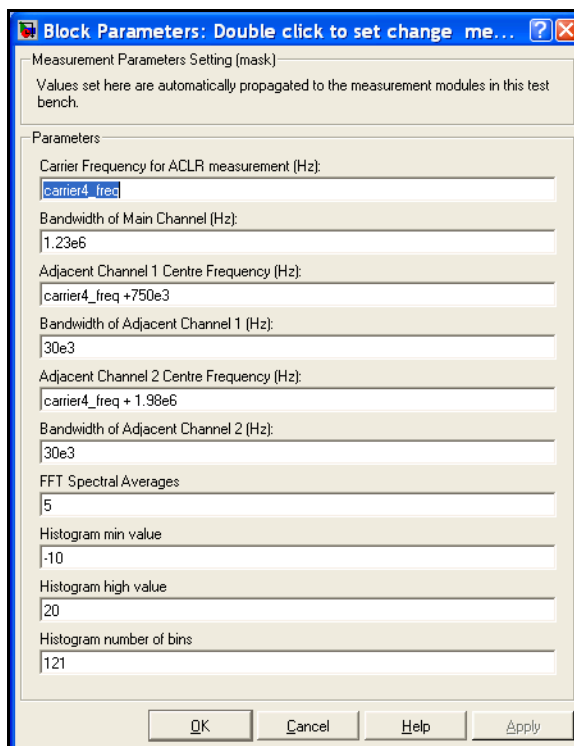
**Figure 11. Measurement Parameters**

Table 4 shows the measurement parameters.

<b>Table 3. Measurement Parameters</b>	
<b>Parameter</b>	<b>Description</b>
Carrier Frequency for ACLR measurement (Hz)	Centre frequency of desired channel for ACLR measurement.
Bandwidth of Main Channel (Hz)	Bandwidth over which to taken the power measurement for the desired channel.
Adjacent Channel 1 Centre Frequency (Hz)	Centre frequency of the first unwanted interferer for ACLR measurement.
Bandwidth of Adjacent Channel 1 Channel (Hz)	Bandwidth over which to take the power measurement for the first unwanted interferer.
Adjacent Channel 2 Centre Frequency (Hz)	Centre frequency of the second unwanted interferer for ACLR measurement.
Bandwidth of Adjacent Channel 2 Channel (Hz)	Bandwidth over which to take the power measurement for the second unwanted interferer.
FFT Spectral Averages	The number of frames of data that are combined for a single plot of the output spectrum. Low numbers cause the scope to update faster but with a less accurate measurement.
Histogram min value	The lowest value for the peak-to-average ratio PDF calculation (dB).
Histogram max value	The highest value for the peak-to-average ratio PDF calculation (dB).
Histogram number of bins	The number of bins in the the peak-to-average ratio PDF calculation. For example, when set to 121 with a maximum value of 20dB and a minimum of –10 dB, the resolution is given by:  $(20 - (-10))/(121 - 1) = 0.25\text{dB}$

### Instantiate the CFR Module in a New Design

To instantiate the CFR module in a new design from the Simulink library, follow these steps:

1. Open the MATLAB and Simulink software.
2. Open the Altera Wireless Library.
3. Open the Radio Card Library.
4. Open the Crest Factor Reduction Library.
5. Open the Peak Windowing Library.
6. Drag the CFR\_peak\_window module into the Simulink project.

To choose the parameters, double click on the CFR block (see Figure 12).

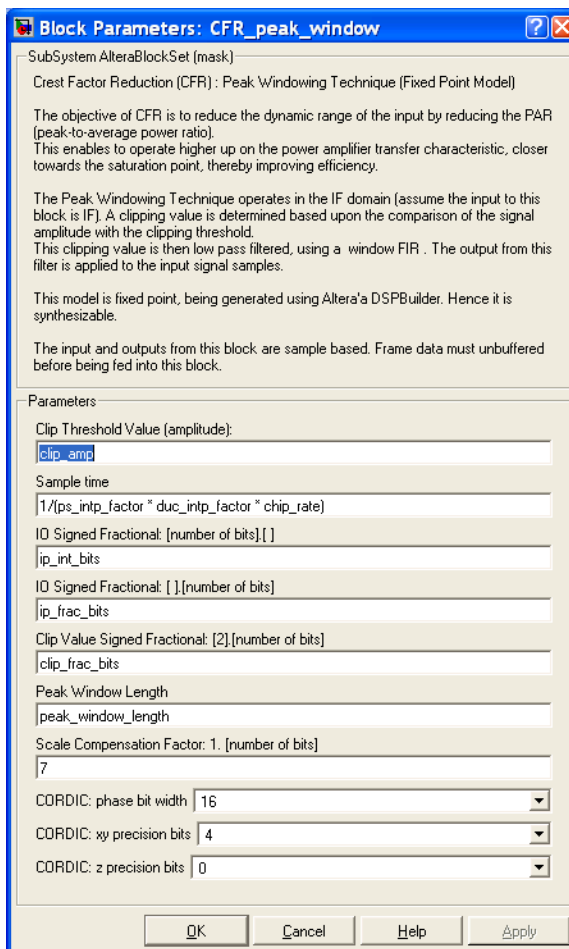
**Figure 12. Parameters**

Table 4 shows the CFR module parameters.

<b>Table 4. Parameters (Part 1 of 2)</b>	
<b>Parameter</b>	<b>Description</b>
Clip threshold value (amplitude)	The value at which the clipping occurs.
Sample time	The sampling interval at CFR input.



**Table 4. Parameters (Part 2 of 2)**

Parameter	Description
IO signed fractional: [number of bits].[ ]	The number of integer bits in the input data and datapath in the design. For example, for one 16-bit signed number less than 1.0 (Q1.15), set to 1.
IO signed fractional: [ ].[number of bit]s	The number of fractional bits in the input data and datapath in the design. For example, for one 16 bit signed number less than 1.0 (Q1.15), set to 15.
Clip Value Signed Fractional: [2].[number of bit]	The number of fractional bits used for the clipping calculation in the design. Two magnitude and sign bits are used by default (the number is $2.0 > \text{value} > -2.0$ ). Typically this value can be less than the input bit width, to reduce complexity. For example, for 11 bits the overall number format is Q2.11.
Peak Window Length	The width of the feed-forward filter impulse response, in samples. The feedback filter impulse response is one-half of this value.
Scale Compensation factor	Scaling for peak window calculation.
CORDIC phase bit width	The size of the calculation for phase in CORDIC.
CORDIC XY precision	The increase in the resolution in the CORDIC magnitude calculation.
CORDIC Z precision bits	The increase in the resolution of the phase calculation in the CORDIC (not required).

### Run the Simulation

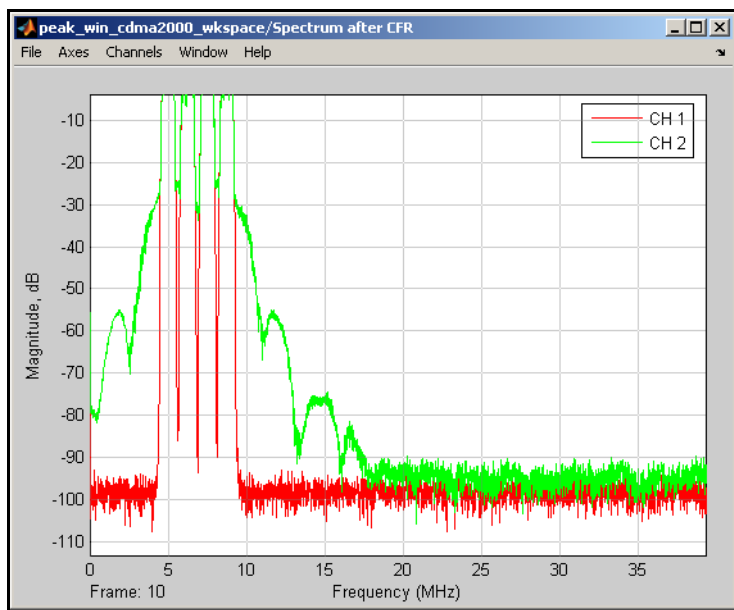
To run the simulation, choose **Start** (Simulation menu).

### Simulation Visualization

Figures 13 and 14 show the frequency spectrum after CFR and after digital downconversion, which shows the spectral leakage into the adjacent channels.



For the testbenches, Altera provides several measurement blocks.

**Figure 13. Upconverted Spectrum Without CFR (CH1) and with CFR (CH2)**

**Figure 14. Baseband Spectrum For All Four Carriers After DDC**

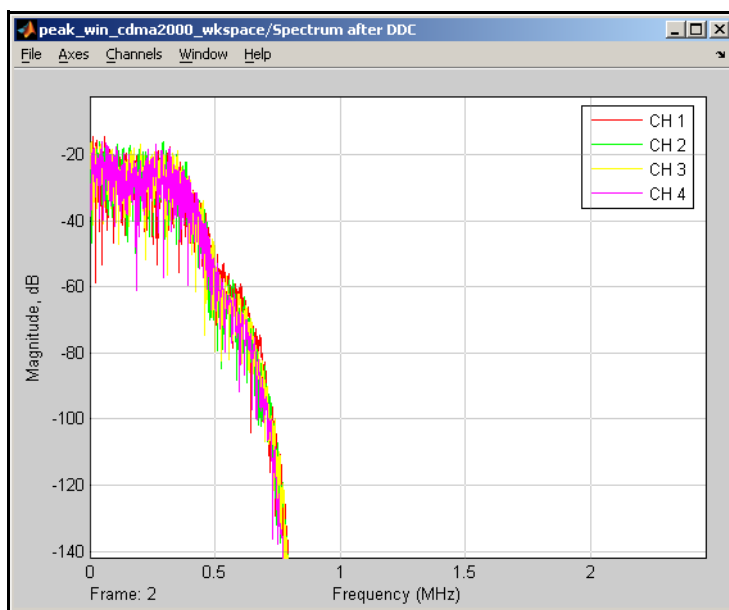
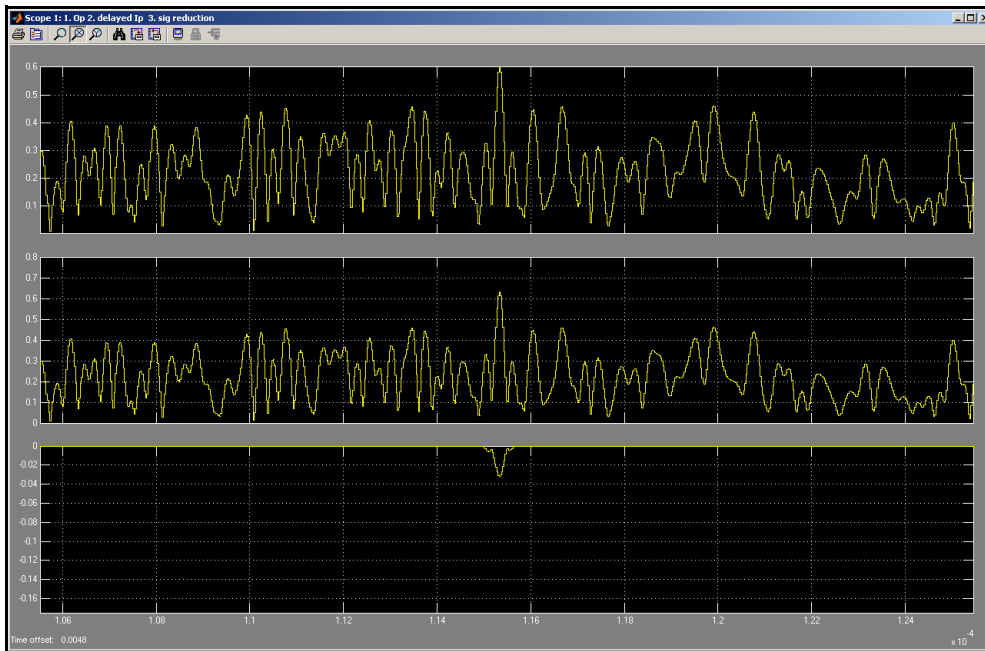


Figure 15 shows the following waveforms:

- Scope 1 is CFR output (magnitude should be less than 0.60), the input signal, and the signal reduction through clipping
- Scope 2 is the delayed input, raw clipping signal (without smoothing) and smoothed clipping signal. These scopes are available for debug purposes only
- Scope 3 is the input to the feedforward FIR filter, the raw FIR input (before negative components are removed), and the output of the feedback FIR filter (which reduces overclipping)

**Figure 15. Waveforms**

## Autogenerate RTL

To autogenerate the RTL for the design, follow these steps:

1. Open the Signal Compiler Module in the Simulink testbench.
2. Click **Convert MDL to VHDL**.
3. Autogenerated VHDL files are in *<install directory>\altera\_wireless\ip\cfr\test\matlab\*.



The design is a mix of Verilog HDL (CORDIC) and VHDL (everything else).

## Synthesize the Design

To synthesize the design, follow these steps:

1. Open SignalCompiler block in the Simulink testbench.

1. Click **Execute steps 1, 2 and 3** to autogenerate RTL, and perform synthesis and fitting.

To modify the synthesis and fitting settings, follow these steps:

1. Change to the `<install dir>\altera_wireless\ip\cfr\test\matlab\` directory.
2. Open the Quartus II software.
3. Open the DSP Builder-generated Quartus II project with the same name as the Simulink model file.
4. Configure the constraints on the design and select the device type (if different from default).
5. Choose **Start Compilation** (Processing menu).

## Perform RTL Simulation

To perform RTL simulation, follow these steps:

6. Open the SignalCompiler block in the Simulink testbench
7. Click the Testbench tab.
8. Turn on **Generate stimuli for VHDL testbench**.
9. Click **OK**.
10. In Simulink choose **Start** (Simulation menu). Input data is logged to files `<install dir>\altera_wireless\ip\cfr\test\matlab\DSP Builder_<Simulink model name>\<signal name>.salt`.

When generating VHDL, DSP Builder also generates a testbench `<install dir>\altera_wireless\ip\cfr\test\matlab\tb_<Simulink model name>.vhd`.

11. Open `tb_<Simulink model name>.tcl`.
12. Replace the following code:

```
puts
"#####"
##### "
puts "# "
puts "#      DSP Builder INFORMATION :"
```



## Performance

Resource usage depends on window length and window type (rounded or triangular). Table 5 and Table 6 show performance for triangular window, length 64.

**Table 5. Performance—Triangular Window, Length 64**

Device	LEs	9 × 9 Multipliers	Memory			f <sub>MAX</sub> (MHz)
			Bits	M4K	M9K	
Cyclone	3754	0	36580	6	—	94
Cyclone II	2823	12	3658	6	—	94
Cyclone III	2801	12	3658	—	6	95

**Table 6. Performance—Triangular Window, Length 64**

Device	Combinational ALUTs	Logic Registers	18 × 18 Multipliers	Memory				f <sub>MAX</sub> (MHz)
				Bits	M4K	M9K	M512	
Stratix II	1922	1867	20	3658	2	—	4	111
Stratix III	1922	1867	24	3658	—	6	—	111

Table 7 and Table 8 show performance for rounded window, length 64..

**Table 7. Performance—Rounded Window, Length 64**

Device	LEs	9 × 9 Multipliers	Memory			f <sub>MAX</sub> (MHz)
			Bits	M4K	M9K	
Cyclone II	2879	12	3664	8	—	100
Cyclone III	2880	12	3664	—	8	107

**Table 8. Performance—Rounded Window, Length 64**

Device	Combinational ALUTs	Logic Registers	18 × 18 Multipliers	Memory				f <sub>MAX</sub> (MHz)
				Bits	M4K	M9K	M512	
Stratix II	2008	1953	20	3664	2	—	6	115
Stratix III	2008	1953	24	3664	—	8	—	114

## Appendix A: Results

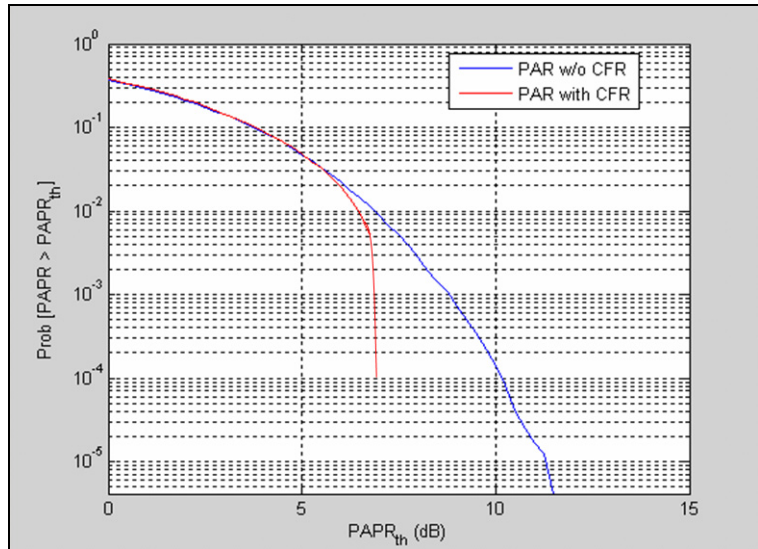
This appendix list results for W-CDMA and CDMA-2000.

## W-CDMA

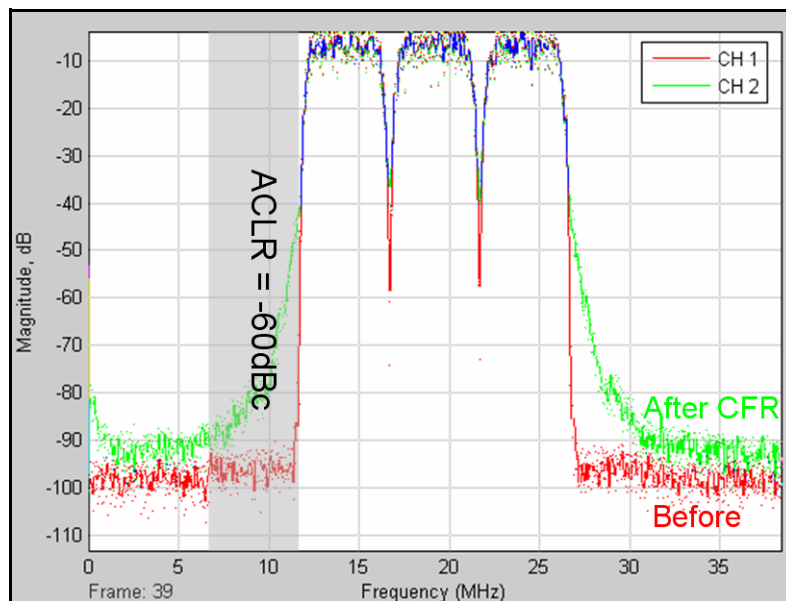
Figures 16 and 17 show the performance for W-CDMA, with three carriers for test case 1. The results have the following conditions:

- EVM = 13.8%
- Output PAR = 6.5dB
- PAR reduction = 4dB at probably  $10^{-4}$

**Figure 16. Complementary Cumulative Density Function (CCDF)**



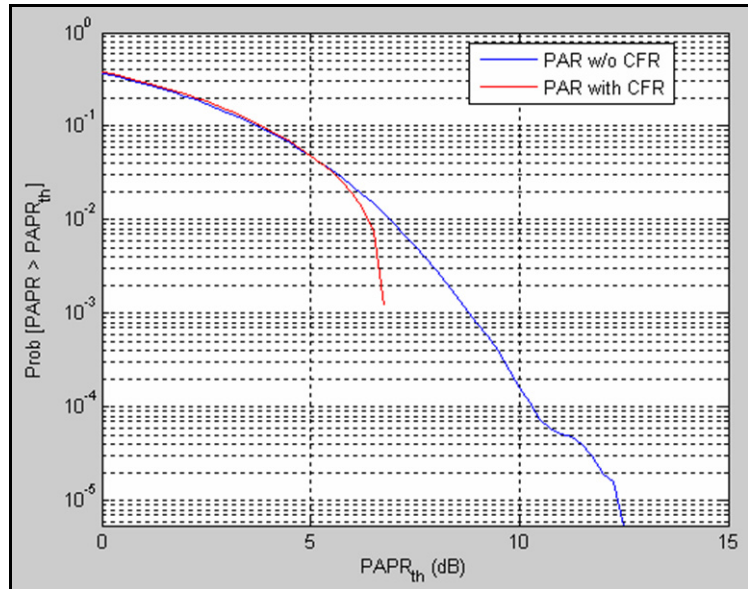


**Figure 17. Spectral Leakage with CFR**

## CDMA-2000

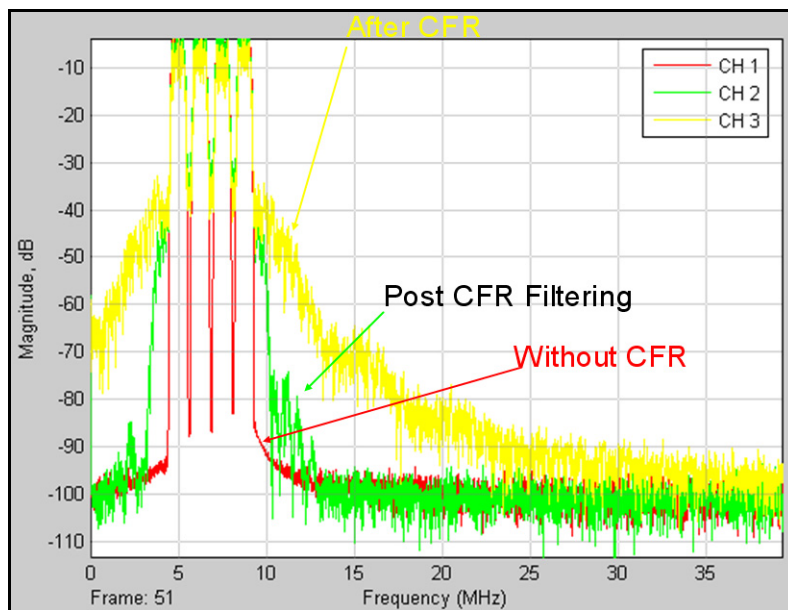
Figure 18 shows indicative performance for CDMA2000, for four carriers, with the following conditions:

- Output PAR at  $10^{-4}$  = 6.7dB
- ACLR after CFR at 750 kHz: -46.0dBc
- 3GPP2 spec: -45dBc
- Channel quality (Rho): 0.9951
- 3GPP2 spec: > 0.985

**Figure 18. CDMA2000 Complementary Cumulative Density Function (CCDF)**

In addition you can apply post-CFR filtering to soften the spectral leakage caused by clipping. Figure 19 shows a filter of approx 10k LEs with the following conditions:

- Output PAR at  $10^{-4} = 6.5\text{dB}$
- ACLR after CFR at 750 kHz :  $-47.5\text{dBc}$
- Minimum Rho: 0.99227

**Figure 19. Spectral Response With & Without Post-CFR Filtering**

## Appendix B: Measurement Definition

This appendix describes the following measurement definitions:

- “PAR” on page 27
- “ACLR” on page 27
- “Error Vector Magnitude” on page 28
- “Rho” on page 28

### PAR

The design calculates the PAR from each sample divided by the mean, as by Agilent. The design records the PAR CCDF for the entire run and the maximum output PAR over the entire 20 ms run. The PAR is the difference between systems with and without CFR and is the maximum over the entire simulation.

### ACLR

The design measures the ACLR at 750 kHz and 1.98 MHz. The ACLR is offset from the highest carrier at 13.69 MHz for systems with CFR. The IF calculation has a resolution bandwidth of 30 kHz. The FFT block size is 8192 LEs, which gives a resolution of 9.6 kHz.

## Error Vector Magnitude

The design calculates the error vector magnitude every slot and records the maximum for the entire run.

## Rho

The 3GPP2 specification gives the following definition for rho:

- $Z_k$  is the  $k$ th sample of the output of the complementary filter
- $R_{0,k}$  is the corresponding sample of the ideal output of the complementary filter.



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